



Quad Low Power Operational Amplifiers

The LM324 series are low–cost, quad operational amplifiers with true differential inputs. They have several distinct advantages over standard operational amplifier types in single supply applications. The quad amplifier can operate at supply voltages as low as 3.0 V or as high as 32 V with quiescent currents about one–fifth of those associated with the MC1741 (on a per amplifier basis). The common mode input range includes the negative supply, thereby eliminating the necessity for external biasing components in many applications. The output voltage range also includes the negative power supply voltage.

- Short Circuited Protected Outputs
- True Differential Input Stage
- Single Supply Operation: 3.0 V to 32 V
- Low Input Bias Currents: 100 nA Maximum (LM324A)
- Four Amplifiers Per Package
- Internally Compensated
- Common Mode Range Extends to Negative Supply
- Industry Standard Pinouts
- ESD Clamps on the Inputs Increase Ruggedness without Affecting Device Operation

QUAD DIFFERENTIAL INPUT OPERATIONAL AMPLIFIERS

SEMICONDUCTOR TECHNICAL DATA





ORDERING INFORMATION

Device	Operating Temperature Range	Package				
LM2902D	$T_{4} = 40^{\circ} t_{2} \pm 105^{\circ}C$	SO-14				
LM2902N	$T_A = -40^\circ \text{ to } +105^\circ \text{C}$	Plastic DIP				
LM224D	T _A = −25° to +85°C	SO-14				
LM224N	IA = -23 to +03 C	Plastic DIP				
LM324AD		SO-14				
LM324AN	$T_{\Delta} = 0^{\circ} \text{ to } +70^{\circ}\text{C}$	Plastic DIP				
LM324D	A - 0 10 + 70 C	SO-14				
LM324N		Plastic DIP				

MAXIMUM RATINGS ($T_A = +25^{\circ}C$, unless otherwise noted.)

Rating	Symbol	LM224 Symbol LM324,A		Unit
Power Supply Voltages Single Supply Split Supplies	V _{CC} V _{CC} , V _{EE}	32 ±16	26 ±13	Vdc
Input Differential Voltage Range (See Note 1)	VIDR	±32	±26	Vdc
Input Common Mode Voltage Range	VICR	-0.3 to 32	-0.3 to 26	Vdc
Output Short Circuit Duration	tSC	Conti	nuous	
Junction Temperature	Тj	15	50	°C
Storage Temperature Range	T _{stg}	–65 to	+150	°C
Operating Ambient Temperature Range	т _А	-25 to +85 0 to +70	-40 to +105	°C

NOTE: 1. Split Power Supplies.

ELECTRICAL CHARACTERISTICS	$(V_{CC} = 5.0 \text{ V}, V_{EE} = \text{GND}, T_A = 25^{\circ}\text{C}, \text{ unless otherwise noted})$
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Characteristics		LM224			LM324A				LM324		LM2902				
	Symbol	Min	Тур	Max	Min	Тур	Мах	Min	Тур	Max	Min	Тур	Max	Unit	
Input Offset Voltage V _{CC} = 5.0 V to 30 V	VIO													mV	
(26 V for LM2902), VICR = 0 V to V _{CC} -1.7 V,															
V_{O} = 1.4 V, R _S = 0 Ω															
T _A = 25°C T _A = T _{high} to T _{Iow} (Note 1)		-	2.0	5.0 7.0	_	2.0	3.0 5.0	_	2.0	7.0 9.0	-	2.0	7.0 10		
Average Temperature Coefficient of Input Offset Voltage	$\Delta V_{IO} / \Delta T$	-	7.0	-	-	7.0	30	-	7.0	-	-	7.0	-	μV/°C	
$T_A = T_{high}$ to T_{low} (Note 1)															
Input Offset Current	IIO	-	3.0	30	-	5.0	30	-	5.0	50	-	5.0	50	nA	
$T_A = T_{high}$ to T_{low} (Note 1)		-		100	-	-	75	-	-	150	-	-	200		
Average Temperature Coefficient of Input Offset Current $T_A = T_{high}$ to T_{low} (Note 1)	ΔΙ _{ΙΟ} /ΔΤ	-	10	-	-	10	300	-	10	-	-	10	-	pA/°C	
Input Bias Current	lin	_	-90	-150	_	-45	-100	_	-90	-250	_	-90	-250	nA	
$T_A = T_{high}$ to T_{low} (Note 1)	IIB	_	-90	-300	_	-43	-200	_	-90	-230 -500	_	-90	-230 -500	ПА	
Input Common Mode Voltage Range (Note 2)	VICR													V	
V _{CC} = 30 V (26 V for LM2902)		0	-	28.3	0	-	28.3	0	-	28.3	0	-	24.3		
$V_{CC} = 30 V (26 V \text{ for LM2902}),$ $T_A = T_{high} \text{ to } T_{low}$		0	-	28	0	-	28	0	-	28	0	-	24		
Differential Input Voltage Range	VIDR	-	-	V _{CC}	V										
Large Signal Open Loop Voltage Gain	AVOL													V/mV	
R _L = 2.0 kΩ, V _{CC} = 15 V, for Large V _O Swing,		50	100	-	25	100	-	25	100	-	25	100	-		
$T_A = T_{high}$ to T_{low} (Note 1)		25	-	-	15	-	-	15	-	-	15	-	-		
$\begin{array}{l} \mbox{Channel Separation} \\ \mbox{10 kHz} \leq f \leq 20 \mbox{ kHz}, \mbox{ Input} \\ \mbox{Referenced} \end{array}$	CS	-	-120	-	-	-120	-	-	-120	-	-	-120	-	dB	
Common Mode Rejection $R_S \le 10 \ k\Omega$	CMR	70	85	-	65	70	-	65	70	-	50	70	-	dB	
Power Supply Rejection	PSR	65	100	-	65	100	-	65	100	-	50	100	-	dB	
Output Voltage-High Limit (T _A = T _{high} to T _{low}) (Note 1)	VOH													V	
$V_{CC} = 5.0 \text{ V}, \text{ R}_{L} = 2.0 \text{ k}\Omega,$ T _A = 25°C		3.3	3.5	-	3.3	3.5	-	3.3	3.5	-	3.3	3.5	-		
$V_{CC} = 30 V (26 V \text{ for LM2902}),$ $R_L = 2.0 \text{ k}\Omega$		26	-	-	26	-	-	26	-	-	22	-	-		
$V_{CC} = 30 \text{ V} (26 \text{ V for LM2902}),$ $R_L = 10 \text{ k}\Omega$		27	28	-	27	28	-	27	28	-	23	24	-		
Output Voltage – Low Limit V_{CC} = 5.0 V, R _L = 10 k Ω , T _A = T _{high} to T _{low} (Note1)	VOL	-	5.0	20	1	5.0	20	1	5.0	20	-	5.0	100	mV	
Output Source Current (V _{ID} = +1.0 V, V _{CC} = 15 V)	IO +													mA	
T _A = 25°C T _A = T _{high} to T _{Iow} (Note 1)		20 10	40 20	-	20 10	40 20	-	20 10	40 20		20 10	40 20			

NOTES: 1. $T_{IOW} = -25^{\circ}C$ for LM224 = 0°C for LM324, A = -40°C for LM2902

 $T_{high} = +85^{\circ}C \text{ for LM224}$ $= +70^{\circ}C \text{ for LM324,A}$ $= +105^{\circ}C \text{ for LM2902}$

2. The input common mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3 V. The upper end of the common mode voltage range is V_{CC} –1.7 V.

Characteristics	Symbol	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Unit
Output Sink Current ($V_{ID} = -1.0 \text{ V}, V_{CC} = 15 \text{ V}$) $T_A = 25^{\circ}\text{C}$	IO –	10	20	-	10	20	_	10	20	-	10	20	-	mA
$T_A = T_{high} \text{ to } T_{IOW} \text{ (Note 1)}$ $(V_{ID} = -1.0 \text{ V}, \text{ V}_O = 200 \text{ mV},$ $T_A = 25^{\circ}\text{C}\text{)}$		5.0 12	8.0 50	-	5.0 12	8.0 50	_	5.0 12	8.0 50	-	5.0 -	8.0 –	-	μΑ
Output Short Circuit to Ground (Note 3)	ISC	-	40	60	-	40	60	-	40	60	-	40	60	mA
$\begin{array}{l} \mbox{Power Supply Current} \\ (T_A = T_{high} \mbox{ to } T_{low}) \mbox{ (Note 1)} \\ \mbox{V}_{CC} = 30 \ \mbox{V} \ (26 \ \mbox{V for LM2902}), \\ \mbox{V}_O = 0 \ \mbox{V}, \ \mbox{R}_L = \infty \end{array}$	ICC	-	_	3.0	_	1.4	3.0	_	_	3.0	_	_	3.0	mA
$V_{CC} = 5.0 \text{ V}, \text{ V}_{O} = 0 \text{ V}, \text{ R}_{L} = \infty$		-	-	1.2	-	0.7	1.2	-	-	1.2	-	-	1.2	

NOTES: 1. $T_{Iow} = -25^{\circ}$ C for LM224 $= 0^{\circ}$ C for LM324, A $= -40^{\circ}$ C for LM320, A $= -40^{\circ}$ C for LM2902 3. Short circuits from the output to V_{CC} can cause excessive heating and eventual destruction. Destructive dissipation can result from simultaneous shorts on all amplifiers.



CIRCUIT DESCRIPTION

The LM324 series is made using four internally compensated, two-stage operational amplifiers. The first stage of each consists of differential input devices Q20 and Q18 with input buffer transistors Q21 and Q17 and the differential to single ended converter Q3 and Q4. The first stage performs not only the first stage gain function but also performs the level shifting and transconductance reduction functions. By reducing the transconductance, a smaller compensation capacitor (only 5.0 pF) can be employed, thus saving chip area. The transconductance reduction is accomplished by splitting the collectors of Q20 and Q18. Another feature of this input stage is that the input common mode range can include the negative supply or ground, in single supply operation, without saturating either the input devices or the differential to single-ended converter. The second stage consists of a standard current source load amplifier stage.



Each amplifier is biased from an internal–voltage regulator which has a low temperature coefficient thus giving each amplifier good temperature characteristics as well as excellent power supply rejection.



Split Supplies





Figure 3. Large–Signal Frequency Response



Figure 4. Small–Signal Voltage Follower Pulse Response (Noninverting)





Figure 6. Input Bias Current versus Power Supply Voltage



Figure 7. Voltage Reference





Figure 9. High Impedance Differential Amplifier



Figure 10. Comparator with Hysteresis



Figure 11. Bi–Quad Filter



Figure 12. Function Generator



Figure 13. Multiple Feedback Bandpass Filter



Given: f_0 = center frequency A(f_0) = gain at center frequency

Choose value fo, C

Then: R3 =
$$\frac{Q}{\pi f_0 C}$$

R1 = $\frac{R3}{2 A(f_0)}$
R2 = $\frac{R1 R3}{4Q^2 R1 - R3}$

For less than 10% error from operational amplifier, $\frac{Q_0 f_0}{BW} < 0.1$ where f₀ and BW are expressed in Hz.

If source impedance varies, filter may be preceded with voltage follower buffer to stabilize filter parameters.

OUTLINE DIMENSIONS



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